



ΕΛΛΗΝΙΚΗ ΔΗΜΟΚΡΑΤΙΑ
ΠΑΝΕΠΙΣΤΗΜΙΟ ΚΡΗΤΗΣ

Ψηφιακή Επεξεργασία Φωνής

Ενότητα 4η: Γραμμική Πρόβλεψη: Ανάλυση και
Σύνθεση

Στυλιανού Ιωάννης

Τμήμα Επιστήμης Υπολογιστών

CS578- SPEECH SIGNAL PROCESSING

LECTURE 4: LINEAR PREDICTION OF SPEECH; ANALYSIS AND SYNTHESIS

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OUTLINE

① TOWARDS LINEAR PREDICTION, LP

② LINEAR PREDICTION

③ ANALYSIS

- Covariance Method
- Autocorrelation Method
- Properties of the Autocorrelation method
- Frequency-Domain Interpretation
- Criterion of goodness
- Comparing Covariance and Autocorrelation

④ SYNTHESIS

⑤ ACKNOWLEDGMENTS

⑥ REFERENCES

TRANSFER FUNCTION FROM THE GLOTTIS TO THE LIPS

- We shown that for voiced speech:

$$\begin{aligned} H(z) &= AG(z)V(z)R(z) \\ &= A \frac{(1 - az^{-1})}{(1 - bz)^2(1 - \sum_{k=1}^N a_k z^{-k})} \end{aligned}$$

- However:

$$1 - az^{-1} = \frac{1}{\sum_{k=0}^{\infty} a_k z^{-k}}, \quad \text{for } |z| > |a|$$

- Then:

$$H(z) = \frac{A}{1 - \sum_{k=1}^p a_k z^{-k}}$$

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PRODUCING SPEECH [1]

Assuming as input to $H(z)$ a train of unit samples, $u_g[n]$, with z-transform $U_g(z)$, then speech, $S(z)$ is given by:

$$H(z) = \frac{S(z)}{U_g(z)} = \frac{A}{1 - \sum_{k=1}^p a_k z^{-k}}$$

or

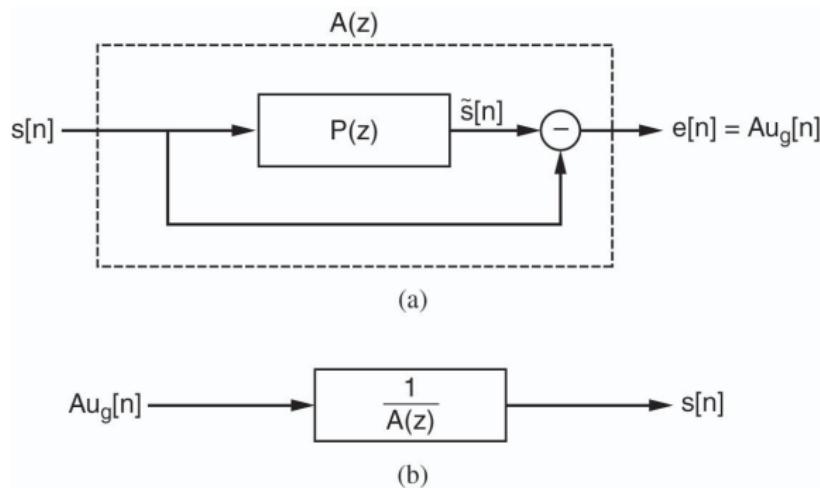
$$S(z) = \sum_{k=1}^p a_k z^{-k} S(z) + A U_g(z)$$

and in time domain:

$$s[n] = \sum_{k=1}^p a_k s[n - k] + A u_g[n]$$

Useful terms: *Linear prediction coefficients, Autoregressive (AR) model/process, Linear prediction analysis*

FILTERING VIEW OF LINEAR PREDICTION



where

$$P(z) = \sum_{k=1}^p a_k z^{-k} \text{ prediction filter}$$

$$A(z) = 1 - P(z) \text{ prediction error filter}$$

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JUSTIFICATION OF LP FOR SPEECH

- If speech is (almost) an AR process, then:

$$s[n] = \sum_{k=1}^p a_k s[n - k] + A u_g[n]$$

- A p th linear predictor, means:

$$\tilde{s}[n] = \sum_{k=1}^p l_k s[n - k]$$

- *Prediction error:*

$$e[n] = s[n] - \tilde{s}[n]$$

- or:

$$e[n] \approx A u_g[n] \quad \text{if } a_k \approx l_k, \forall k$$

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ERROR MINIMIZATION

- Over all time we wish to minimize the mean-squared prediction error:

$$E = \sum_{m=-\infty}^{\infty} (s[m] - \tilde{s}[m])^2$$

- Prediction error in the vicinity of n :

$$E_n = \sum_{m=n-M}^{n+M} (s[m] - \tilde{s}[m])^2$$

- Prediction interval:* $[n - M, n + M]$

-

$$E_n = \sum_{m=-\infty}^{\infty} e_n^2[m]$$

where

$$e_n[m] = s_n[m] - \sum_{k=1}^p l_k s_n[m-k], \quad n - M \leq m \leq n + M$$

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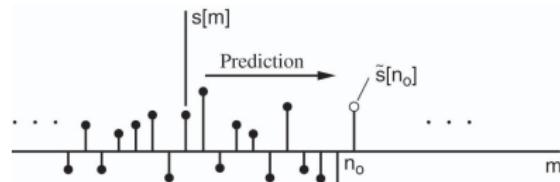
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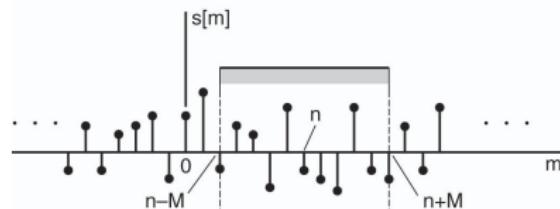
COVARIANCE METHOD

- Samples outside the prediction error interval are NOT zero
- Minimization of the mean-squared error in the prediction error interval

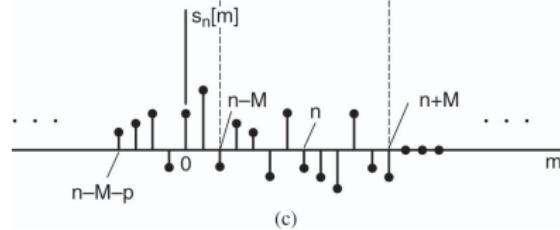
SHORT-TIME SEQUENCES: COVARIANCE



(a)



(b)



(c)

COVARIANCE METHOD: FORMULATION

- In matrix notation

$$\mathbf{e}_n^{(2M+1 \times 1)} = \mathbf{s}_n^{(2M+1 \times 1)} - \mathbf{S}_n^{(2M+1 \times p)} \mathbf{I}^{(p \times 1)}$$

- Mean-squared error

$$\mathbf{e}_n^T \mathbf{e}_n = \mathbf{s}_n^T \mathbf{s}_n - 2\mathbf{s}_n^T \mathbf{S}_n \mathbf{I} + \mathbf{I}^T \mathbf{S}_n^T \mathbf{S}_n \mathbf{I}$$

- Solution:

$$\mathbf{I} = \left(\mathbf{S}_n^T \mathbf{S}_n \right)^{-1} \mathbf{S}_n^T \mathbf{s}_n$$

- Same solution by considering the *Projection Theorem*:

$$\mathbf{S}_n^T \mathbf{e}_n = 0$$

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$$\mathbf{e}_n^{(2M+1 \times 1)} = \mathbf{s}_n^{(2M+1 \times 1)} - \mathbf{S}_n^{(2M+1 \times p)} \mathbf{l}^{(p \times 1)}$$

- Mean-squared error

$$\mathbf{e}_n^T \mathbf{e}_n = \mathbf{s}_n^T \mathbf{s}_n - 2\mathbf{s}_n^T \mathbf{S}_n \mathbf{l} + \mathbf{l}^T \mathbf{S}_n^T \mathbf{S}_n \mathbf{l}$$

- Solution:

$$\mathbf{l} = \left(\mathbf{S}_n^T \mathbf{S}_n \right)^{-1} \mathbf{S}_n^T \mathbf{s}_n$$

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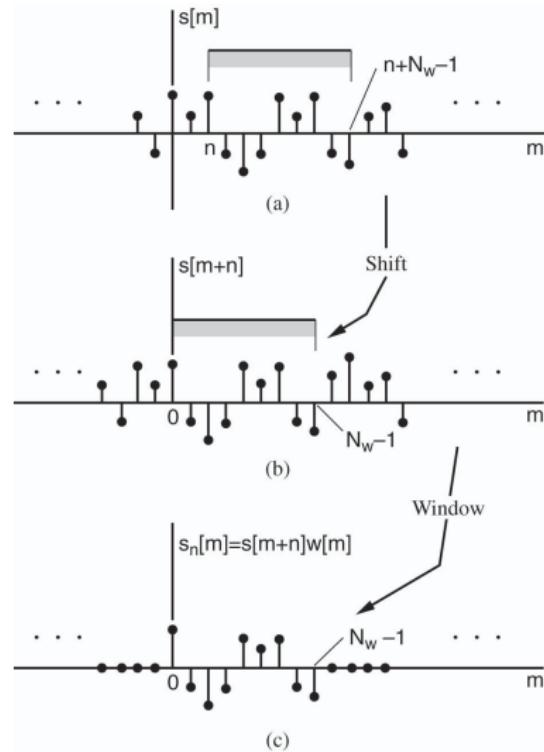
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AUTOCORRELATION METHOD

- Samples outside the prediction error interval are all zero
- Minimization of the mean-squared error in $\pm\infty$

SHORT-TIME SEQUENCES: AUTOCORRELATION



AUTOCORRELATION METHOD: FORMULATION

- Error is nonzero in the interval $[0, N_w + p - 1]$:

$$E_n = \sum_{m=0}^{N_w+p-1} e_n^2[m]$$

- Normal equations:

$$\sum_{k=1}^p l_k \Phi_n[i, k] = \Phi_n[i, 0], \quad i = 1, 2, 3, \dots, p$$

where

$$\Phi_n[i, k] = \sum_{m=0}^{N_w+p-1} s_n[m-i] s_n[m-k], \quad 1 \leq i \leq p, \quad 0 \leq k \leq p$$

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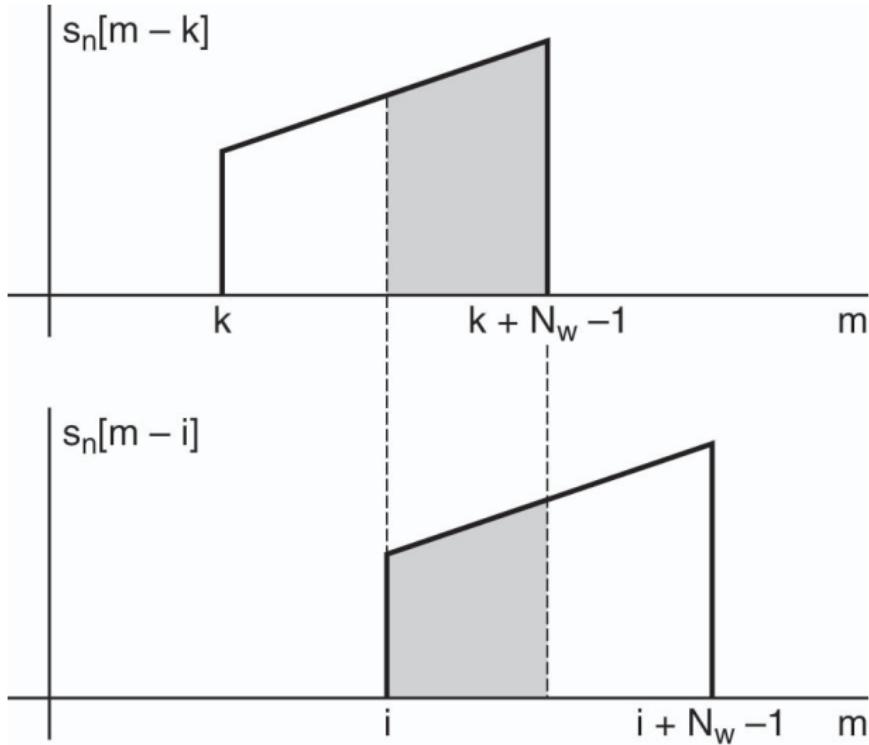
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CONSTRUCTING THE AUTOCORRELATION FUNCTION



USING THE AUTOCORRELATION FUNCTION

- by denoting:

$$r_n[i - k] = \Phi_n[i, k]$$

- Then:

$$\sum_{k=1}^p (p l_k r_n[i - k] = r_n[i], \quad 1 \leq i \leq p$$

- In matrix notation:

$$\mathbf{R}_n^{(p \times p)} \mathbf{l}^{(p \times 1)} = \mathbf{r}_n^{(p \times 1)}$$

- Or (Toeplitz matrix):

$$\begin{bmatrix} r_n[0] & r_n[1] & \cdots & r_n[p-1] \\ r_n[1] & r_n[0] & \cdots & r_n[p-2] \\ \vdots & \vdots & \ddots & \vdots \\ r_n[p-1] & r_n[p-2] & \cdots & r_n[0] \end{bmatrix} \begin{bmatrix} l_1 \\ l_2 \\ \vdots \\ l_p \end{bmatrix} = \begin{bmatrix} r_n[1] \\ r_n[2] \\ \vdots \\ r_n[p] \end{bmatrix}$$

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LEVINSON RECURSION

▷ Build an order $i + 1$ solution from an order i solution until the desired order p is reached:

- Initial step:

$$l_0^0 = 0, \quad E^0 = r[0]$$

- Step 1: Compute the *partial correlation coefficients*

$$k_i = \frac{r[i] - \sum_{j=1}^{i-1} l_j^{i-1} r[i-j]}{E^{i-1}}$$

- Step 2: Update prediction coefficients, l

$$\begin{aligned} l_i^i &= k_i \\ l_j^i &= l_j^{i-1} - k_i l_{i-j}^{i-1}, \quad 1 \leq j \leq i-1 \end{aligned}$$

- Step 3: Update the minimum squared prediction error

$$E^i = (1 - k_i^2)E^{i-1}$$

- Step 4: Repeat steps 1 to 3 for $i = 1, 2, \dots, p$

- Final Step: at p th step, compute the optimal predictor coefficients, l_j^* ,

$$l_j^* = l_j^p, \quad 1 \leq j \leq p$$

LOSSLESS TUBE MODEL AND LINEAR PREDICTION

There is a strong resemblance to the recursions in the lossless tube model and in the Autocorrelation Method for Linear Prediction:

- Transfer functions:

$$V(z) = \frac{A}{D(z)} \quad D(z) = 1 - \sum_{k=1}^N l_k z^{-k}$$

$$H(z) = \frac{A}{A(z)} \quad A(z) = 1 - \sum_{k=1}^p l_k z^{-k}$$

- Recursions:

$D_0(z) = 1$	$A^0(z) = 1$
For $k = 1, 2, \dots, N$	For $i = 1, 2, \dots, p$
$D_k(z) = D_{k-1}(z) + r_k z^{-k} D_{k-1}(z^{-1})$	$A^i(z) = A^{i-1}(z) - k_i z^{-i} A^{i-1}(z^{-1})$
$D(z) = D_N(z)$	$A(z) = A_p(z)$

- Identical recursions if: $k_i = -r_i = -\frac{A_{i+1}-A_i}{A_{i+1}+A_i}$

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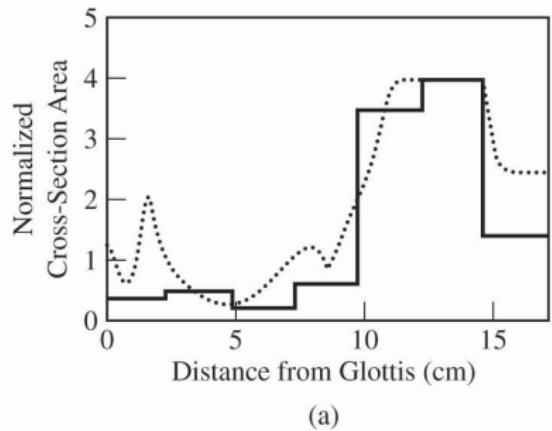
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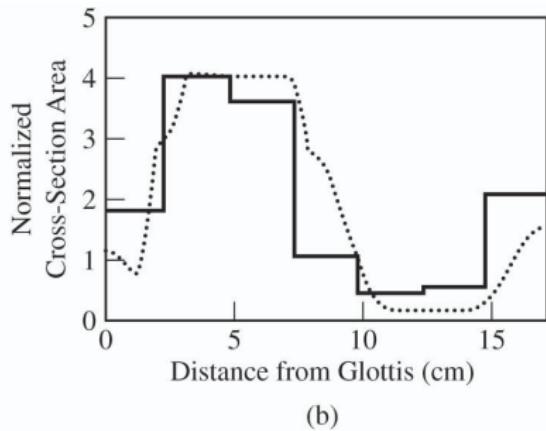
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ESTIMATING THE VOCAL TRACT AREA FUNCTIONS VIA THE AUTOCORRELATION METHOD



(a)



(b)

PROPERTIES OF THE AUTOCORRELATION METHOD

- $|k_i| < 1, \forall i$
- $H(z)$ is a minimum phase system (*stability*)
- Flip all maximum-phase poles inside the unit circle to their conjugate reciprocal locations
- One-to-One correspondence: $k_i \Leftrightarrow l_i, l_i \Leftrightarrow r_n[i]$.

$$\begin{aligned} k_i &= l_i^i \\ l_j^{i-1} &= \frac{l_j^i + k_i l_{i-j}^i}{1 - k_i^2} \end{aligned}$$

- *Autocorrelation matching:* If, $H(z)$ is an p th all-pole minimum phase system, and if $r_h[0] = r_n[0]$, then:

$$r_h[\tau] = r_n[\tau], \text{ for } |\tau| \leq p$$

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- *Autocorrelation matching:* If, $H(z)$ is an p th all-pole minimum phase system, and if $r_h[0] = r_n[0]$, then:

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PROPERTIES OF THE AUTOCORRELATION METHOD

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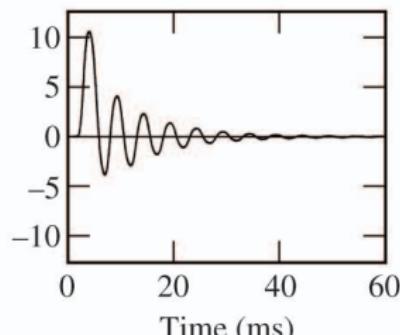
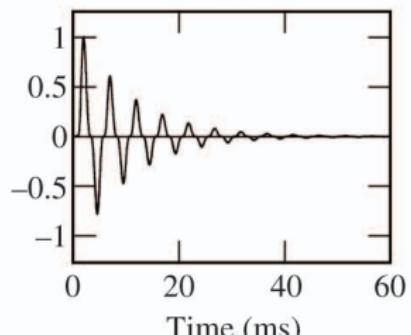
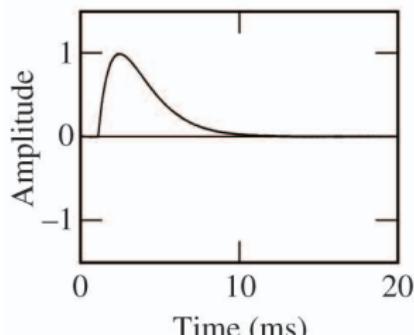
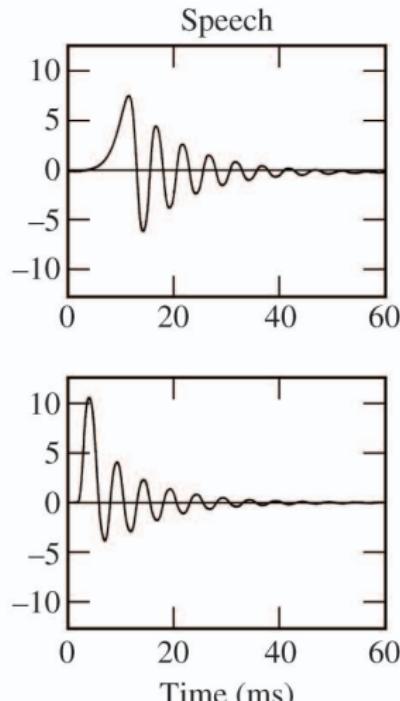
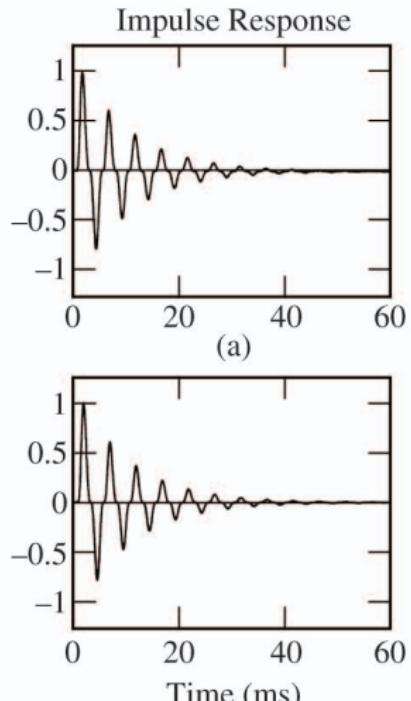
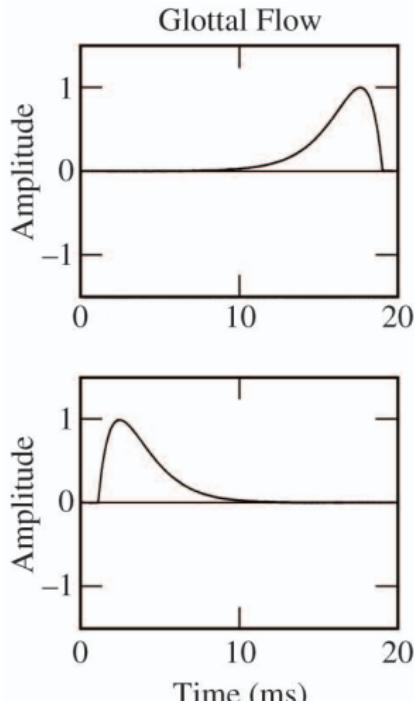
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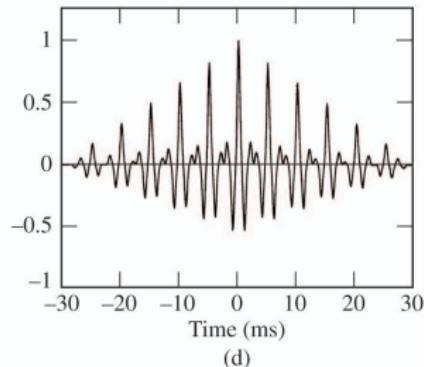
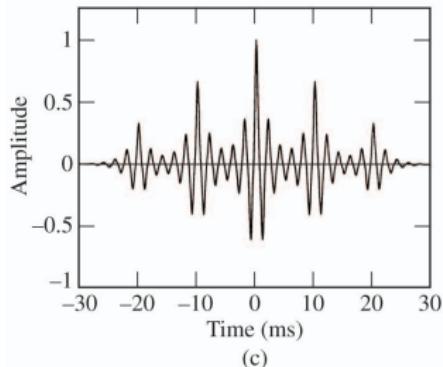
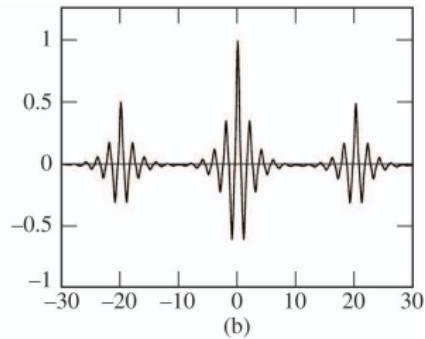
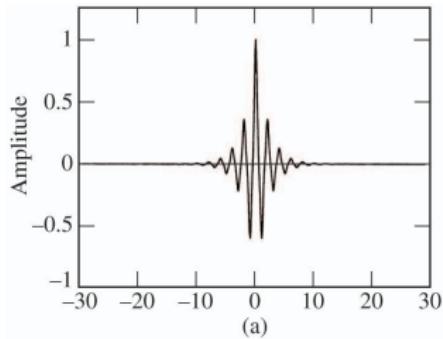
CONSEQUENCE I

- ▷ Flip all maximum-phase poles inside the unit circle to their conjugate reciprocal locations



CONSEQUENCE II

▷ Autocorrelation matching



CONSEQUENCE III

- ▷ Autocorrelation matching:

$$A^2 = r_h[0] - \sum_{k=1}^p l_k r_h[k]$$

or

$$A^2 = r_n[0] - \sum_{k=1}^p l_k r_n[k] = E_n$$

ESTIMATIONS IN THE FREQUENCY DOMAIN

- Let $|S(\omega)|$ be the magnitude spectrum of speech and $H(\omega) = A/A(\omega)$ be an all-pole model
- Define a frequency-domain error function

$$I = \frac{1}{2\pi} \int_{-\pi}^{\pi} [e^{Q(\omega)} - Q(\omega) - 1] d\omega$$

where

$$Q(\omega) = \log |S(\omega)|^2 - \log |H(\omega)|^2 = \log \left| \frac{E(\omega)}{A} \right|^2$$

- Minimizing I over the linear prediction coefficients, results in the minimization of:

$$\int_{-\pi}^{\pi} |E(\omega)|^2 d\omega$$

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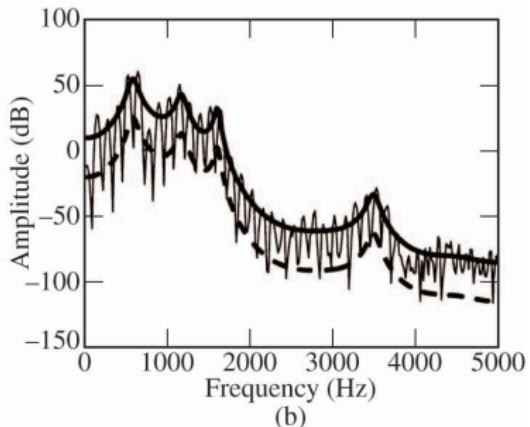
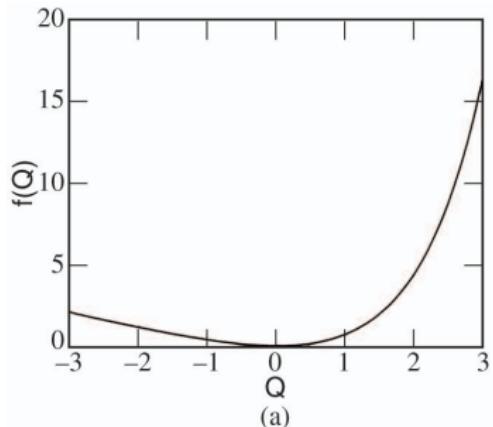
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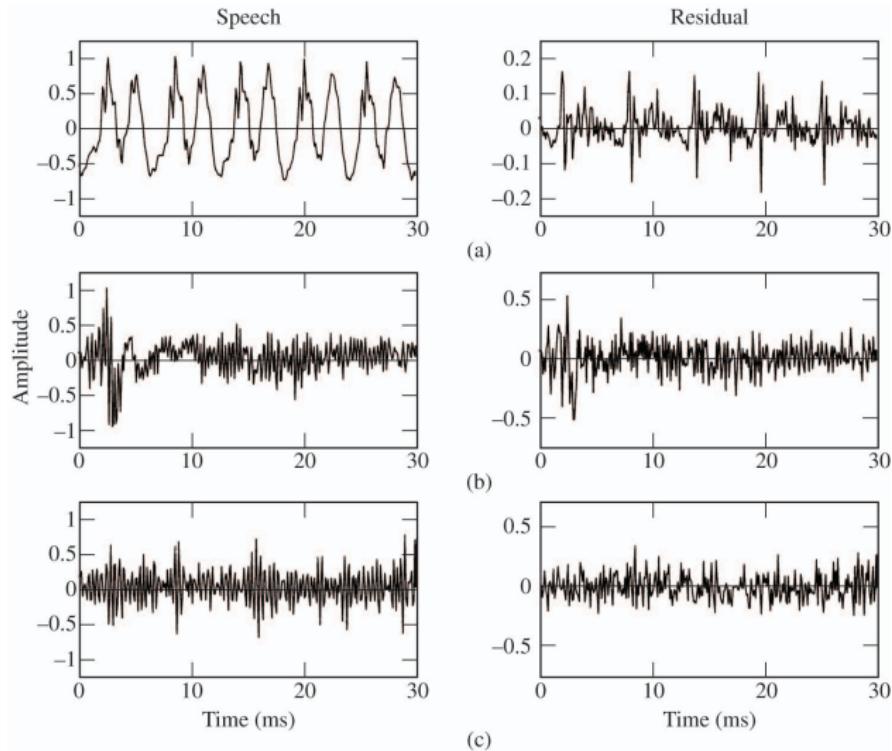
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FAVORING SPECTRAL PEAKS

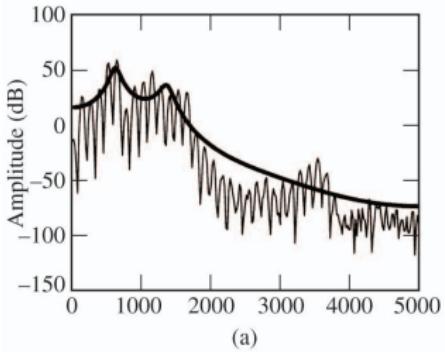
A note on $f(Q) = e^Q - Q - 1$



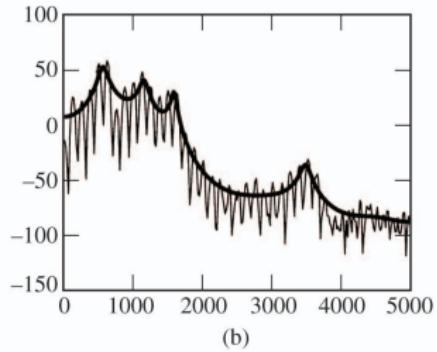
TIME-DOMAIN



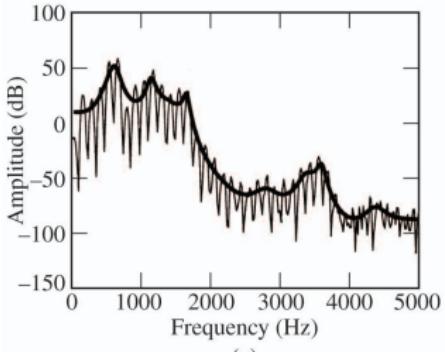
FREQUENCY-DOMAIN: VOICED



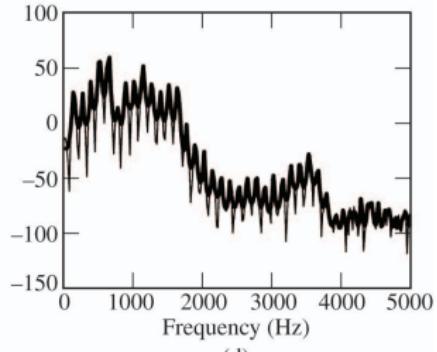
(a)



(b)

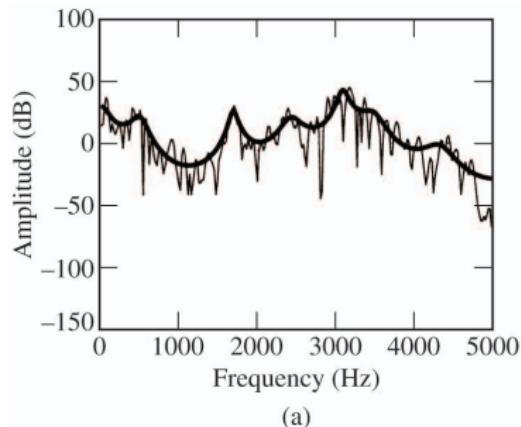


(c)

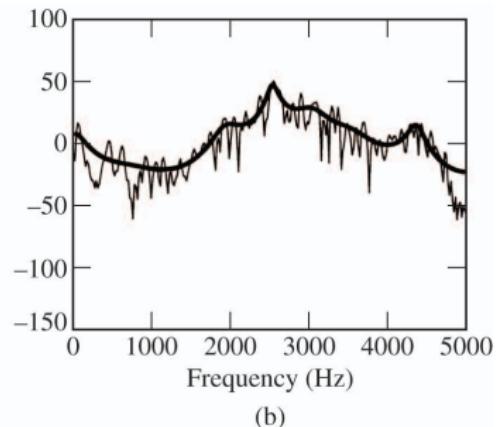


(d)

FREQUENCY-DOMAIN: UNVOICED



(a)



(b)

COMPARING COVARIANCE AND AUTOCORRELATION

- Simple test of estimation

$$s[n] = a^n u[n] \star \delta[n]$$

- Stability issues
- Sensitivity, pitch-synchronous analysis

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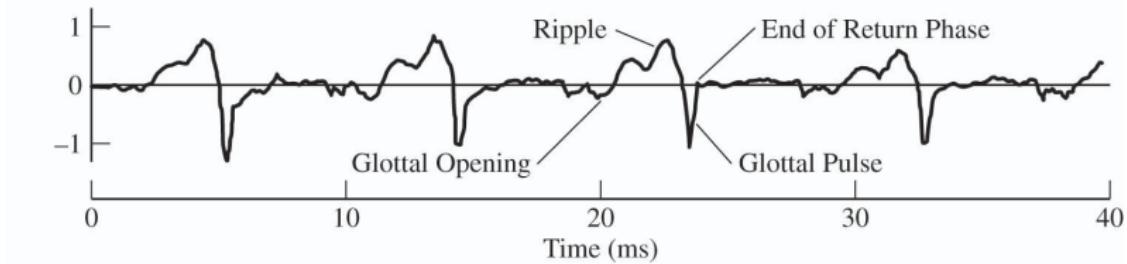
SENSITIVITY, PITCH-SYNCHRONOUS ANALYSIS



(a)



(b)



(c)

OUTLINE

① TOWARDS LINEAR PREDICTION, LP

② LINEAR PREDICTION

③ ANALYSIS

- Covariance Method
- Autocorrelation Method
- Properties of the Autocorrelation method
- Frequency-Domain Interpretation
- Criterion of goodness
- Comparing Covariance and Autocorrelation

④ SYNTHESIS

⑤ ACKNOWLEDGMENTS

⑥ REFERENCES

SYNTHESIS

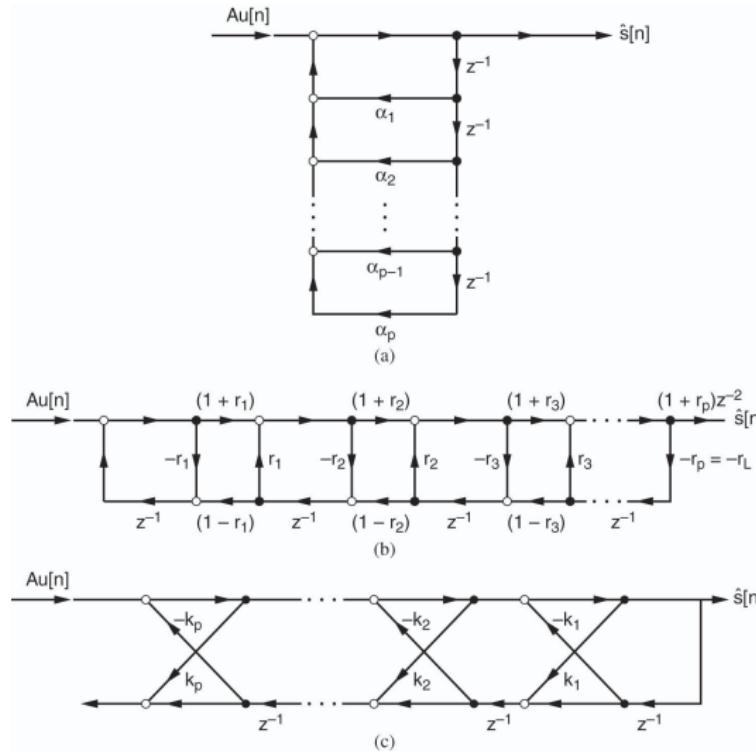
The synthesized speech is:

$$s[n] = \sum_{k=1}^p l_k s[n - k] + A u[n]$$

where $u[n]$ could be:

- A periodic impulse train
- An impulse
- White noise

SYNTHESIS STRUCTURE



CONSIDER ...

- Window duration
- Frame interval (frame rate)
- Model order
- Voiced/unvoiced state and pitch estimation
- Synthesis structure

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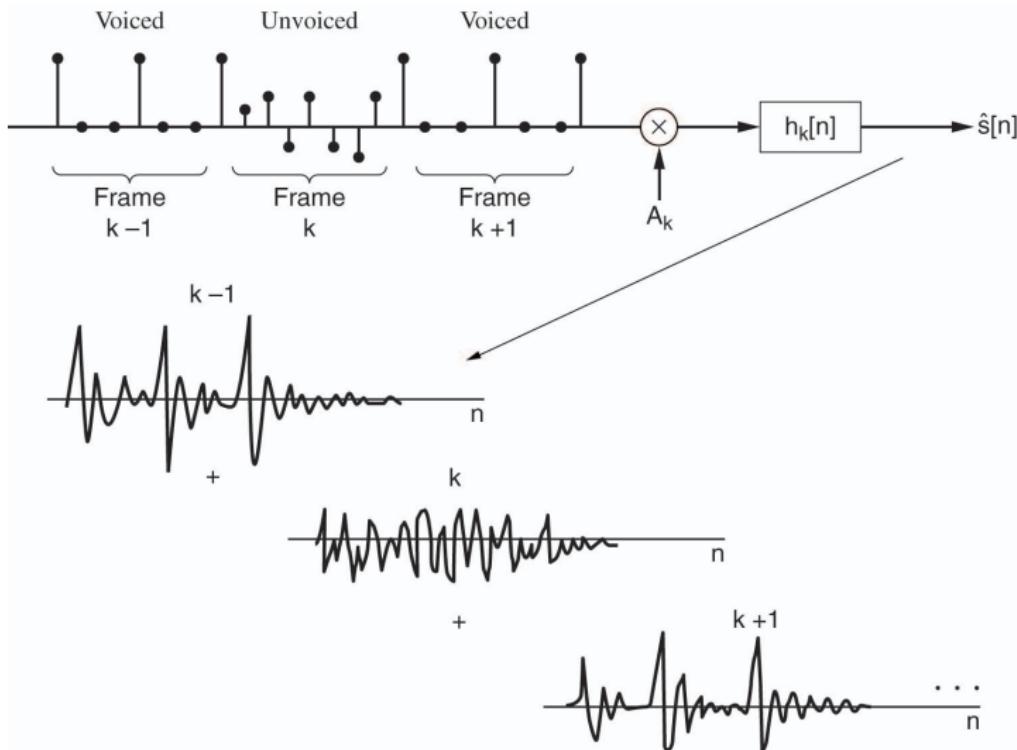
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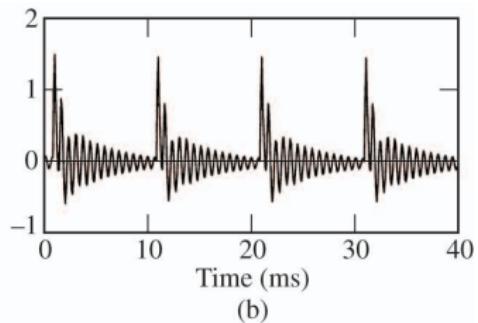
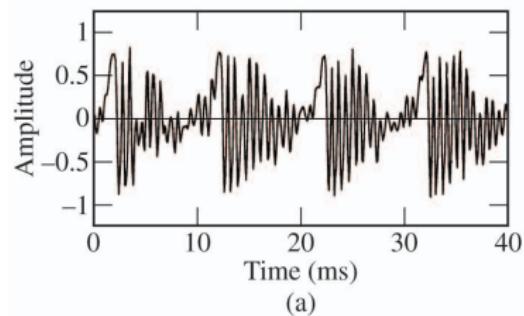
CONSIDER ...

- Window duration
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OVERLAP AND ADD, OLA



SPEECH RECONSTRUCTION EXAMPLE



HOW DOES IT SOUND ...

- /a/



- /e/



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ACKNOWLEDGMENTS

Most, if not all, figures in this lecture are coming from the book:

T. F. Quatieri: Discrete-Time Speech Signal Processing,
principles and practice
2002, Prentice Hall

and have been used after permission from Prentice Hall

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J. Makhoul, "Linear Prediction: A Tutorial Review," *Proceedings of the IEEE*, vol. 63, pp. 561–580, April 1975.

Τέλος Ενότητας



Με τη συγχρηματοδότηση της Ελλάδας και της Ευρωπαϊκής Ένωσης

Χρηματοδότηση

- Το παρόν εκπαιδευτικό υλικό έχει αναπτυχθεί στα πλαίσια του εκπαιδευτικού έργου του διδάσκοντα.
- Το έργο «**Ανοικτά Ακαδημαϊκά Μαθήματα στο Πανεπιστήμιο Κρήτης**» έχει χρηματοδοτήσει μόνο τη αναδιαμόρφωση του εκπαιδευτικού υλικού.
- Το έργο υλοποιείται στο πλαίσιο του Επιχειρησιακού Προγράμματος «Εκπαίδευση και Δια Βίου Μάθηση» και συγχρηματοδοτείται από την Ευρωπαϊκή Ένωση (Ευρωπαϊκό Κοινωνικό Ταμείο) και από εθνικούς πόρους.



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- Ως Μη Εμπορική ορίζεται η χρήση:**
 - που δεν περιλαμβάνει άμεσο ή έμμεσο οικονομικό όφελος από την χρήση του έργου, για το διανομέα του έργου και αδειοδόχο
 - που δεν περιλαμβάνει οικονομική συναλλαγή ως προϋπόθεση για τη χρήση ή πρόσβαση στο έργο
 - που δεν προσπορίζει στο διανομέα του έργου και αδειοδόχο έμμεσο οικονομικό όφελος (π.χ. διαφημίσεις) από την προβολή του έργου σε διαδικτυακό τόπο
- Ο δικαιούχος μπορεί να παρέχει στον αδειοδόχο ξεχωριστή άδεια να χρησιμοποιεί το έργο για εμπορική χρήση, εφόσον αυτό του ζητηθεί.

Σημείωμα Αναφοράς

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Διατήρηση Σημειωμάτων

Οποιαδήποτε αναπαραγωγή ή διασκευή του υλικού θα πρέπει να συμπεριλαμβάνει:

- το Σημείωμα Αναφοράς
- το Σημείωμα Αδειοδότησης
- τη δήλωση Διατήρησης Σημειωμάτων
- το Σημείωμα Χρήσης Έργων Τρίτων (εφόσον υπάρχει)

μαζί με τους συνοδευόμενους υπερσυνδέσμους.

Σημείωμα Χρήσης Έργων Τρίτων

Το Έργο αυτό κάνει χρήση των ακόλουθων έργων:

Εικόνες/Σχήματα/Διαγράμματα/Φωτογραφίες

Εικόνες/σχήματα/διαγράμματα/φωτογραφίες που περιέχονται σε αυτό το αρχείο προέρχονται από το βιβλίο:

Τίτλος: *Discrete-time Speech Signal Processing: Principles and Practice*

Prentice-Hall signal processing series, ISSN 1050-2769

Συγγραφέας: Thomas F. Quatieri

Εκδότης: Prentice Hall PTR, 2002

ISBN: 013242942X, 9780132429429

Μέγεθος: 781 σελίδες

και αναπαράγονται μετά από άδεια του εκδότη.